

# S-1131 Series

www.ablic.com

## HIGH RIPPLE-REJECTION AND LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR

© ABLIC Inc., 2002-2014

Rev.4.1\_02

The S-1131 Series is a positive voltage regulator with a low dropout voltage, high-accuracy output voltage, and low current consumption developed based on CMOS technology.

A built-in low on-resistance transistor provides a low dropout voltage and large output current, and a built-in overcurrent protection circuit prevents the load current from exceeding the current capacitance of the output transistor. An ON/OFF circuit ensures a long battery life, and small SOT-89-3, SOT-89-5 and 6-Pin HSON(A) packages realize high-density mounting.

## Features

- Output voltage:
- Output voltage accuracy:
- Dropout voltage:
- Current consumption:
- Output current:
- Ripple rejection:
- Built-in overcurrent protection circuit:
- Built-in ON/OFF circuit:
- Operation temperature range:
- Lead-free, Sn 100%, halogen-free<sup>\*2</sup>

1.5 V to 5.5 V, selectable in 0.1 V step  $\pm 1.0\%$ 250 mV typ. (3.0 V output product,  $I_{OUT} = 100$  mA) During operation: 35 µA typ., 65 µA max. During power-off: 0.1 µA typ., 1.0 µA max. Possible to output 300 mA ( $V_{IN} \ge V_{OUT(S)} + 1.0$  V)<sup>\*1</sup> 70 dB typ. (f = 1.0 kHz) Limits overcurrent of output transistor. Ensures long battery life. Ta = -40°C to +85°C

- \*1. Attention should be paid to the power dissipation of the package when the output current is large.
- \*2. Refer to "
  Product Name Structure" for details.

## Applications

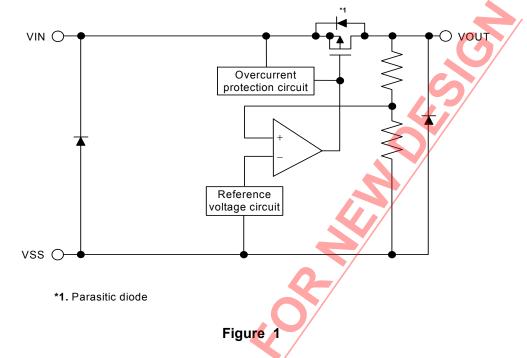
- Constant-voltage power supply for DVD and CD-ROM drive
- Constant-voltage power supply for battery-powered device
- Constant-voltage power supply for personal communication device
- Constant-voltage power supply for note book PC

#### Packages

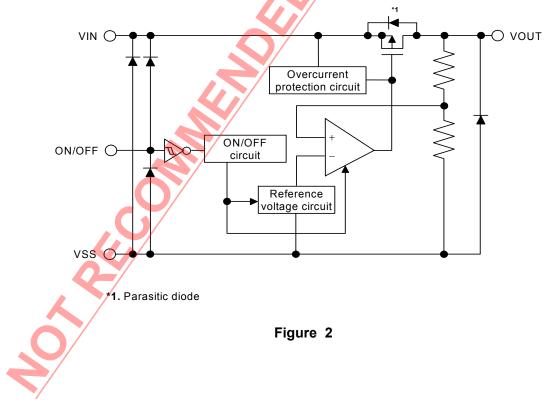
- SOT-89-3
- SOT-89-5
- 6-Pin HSON(A)

## Block Diagrams

(1) Without ON/OFF circuit (Package: SOT-89-3)



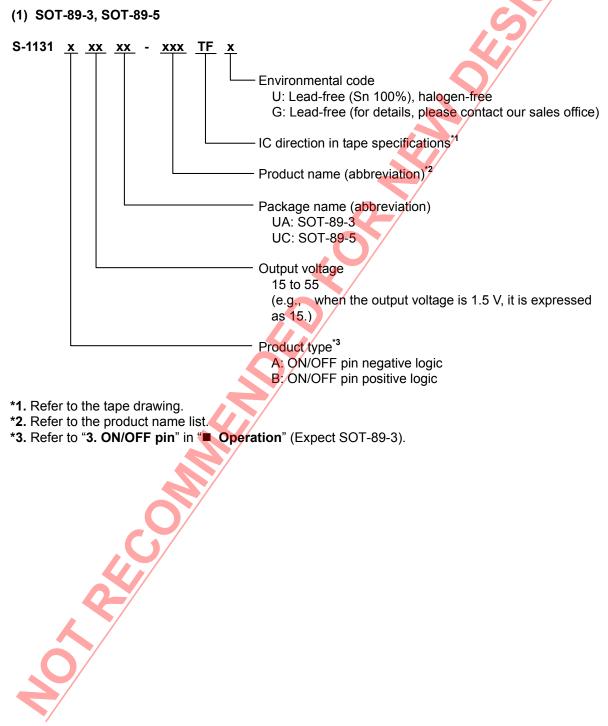
(2) With ON/OFF circuit (Package: SOT-89-5, 6-Pin HSON(A))



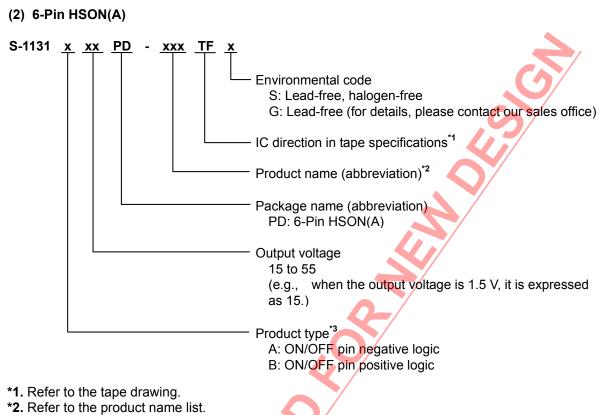
#### Product Name Structure

• Users can select the product type, output voltage, and package type for the S-1131 Series. Refer to "1. **Product name**" regarding the contents of product name, "2. **Packages**" regarding the package drawings, "3. **Product name list**" regarding details of the product name.

#### 1. Product name



## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR S-1131 Series Rev.4.1\_02



\*3. Refer to "3. ON/OFF pin" in "■ Operation"

#### 2. Packages

Dookogo Nomo		Drawing Code	
Package Name	Package	Таре	Reel
SOT-89-3	UP003-A-P-SD	UP003-A-C-SD	UP003-A-R-SD
SOT-89-5	UP005-A-P-SD	UP005-A-C-SD	UP005-A-R-SD
6-Pin HSON(A)	PD006-A-P-SD	PD006-A-C-SD	PD006-A-R-SD

## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR Rev.4.1\_02 S-1131 Series

#### 3. Product name list

		/

Output voltage	SOT-89-3	SOT-89-5	6-Pin HSON(A)
1.5V±1.0%	S-1131B15UA-N4ATFx	S-1131B15UC-N4ATFx	S-1131B15PD-N4ATFz
1.6V±1.0%	S-1131B16UA-N4BTFx	S-1131B16UC-N4BTFx	S-1131B16PD-N4BTFz
1.7V±1.0%	S-1131B17UA-N4CTFx	S-1131B17UC-N4CTFx	S-1131B17PD-N4CTFz
1.8V±1.0%	S-1131B18UA-N4DTFx	S-1131B18UC-N4DTFx	S-1131B18PD-N4DTFz
1.9V±1.0%	S-1131B19UA-N4ETFx	S-1131B19UC-N4ETFx	S-1131B19PD-N4ETFz
2.0V±1.0%	S-1131B20UA-N4FTFx	S-1131B20UC-N4FTFx	S-1131B20PD-N4FTFz
2.1V±1.0%	S-1131B21UA-N4GTFx	S-1131B21UC-N4GTFx	S-1131B21PD-N4GTFz
2.2V±1.0%	S-1131B22UA-N4HTFx	S-1131B22UC-N4HTFx	S-1131B22PD-N4HTFz
2.3V±1.0%	S-1131B23UA-N4ITFx	S-1131B23UC-N4ITEx	S-1131B23PD-N4ITFz
2.4V±1.0%	S-1131B24UA-N4JTFx	S-1131B24UC-N4JTFx	S-1131B24PD-N4JTFz
2.5V±1.0%	S-1131B25UA-N4KTFx	S-1131B25UC-N4KTFx	S-1131B25PD-N4KTFz
2.6V±1.0%	S-1131B26UA-N4LTFx	S-1131B26UC-N4LTFx	S-1131B26PD-N4LTFz
2.7V±1.0%	S-1131B27UA-N4MTFx	S-1131B27UC-N4MTFx	S-1131B27PD-N4MTFz
2.8V±1.0%	S-1131B28UA-N4NTFx	S-1131B28UC-N4NTFx	S-1131B28PD-N4NTFz
2.9V±1.0%	S-1131B29UA-N4OTFx	S-1131B29UC-N4OTFx	S-1131B29PD-N4OTFz
3.0V±1.0%	S-1131B30UA-N4PTFx	S-1131B30UC-N4PTFx	S-1131B30PD-N4PTFz
3.1V±1.0%	S-1131B31UA-N4QTFx	S-1131B31UC-N4QTFx	S-1131B31PD-N4QTFz
3.2V±1.0%	S-1131B32UA-N4RTFx	S-1131B32UC-N4RTFx	S-1131B32PD-N4RTFz
3.3V±1.0%	S-1131B33UA-N4STFx	S-1131B33UC-N4STFx	S-1131B33PD-N4STFz
3.4V±1.0%	S-1131B34UA-N4TTFx	S-1131B34UC-N4TTFx	S-1131B34PD-N4TTFz
3.5V±1.0%	S-1131B35UA-N4UTFx 🚽	S-1131B35UC-N4UTFx	S-1131B35PD-N4UTFz
3.6V±1.0%	S-1131B36UA-N4VTF	S-1131B36UC-N4VTFx	S-1131B36PD-N4VTFz
3.7V±1.0%	S-1131B37UA-N4WTFx	S-1131B37UC-N4WTFx	S-1131B37PD-N4WTFz
3.8V±1.0%	S-1131B38UA-N4XTFx	S-1131B38UC-N4XTFx	S-1131B38PD-N4XTFz
3.9V±1.0%	S-1131B39UA-N4YTFx	S-1131B39UC-N4YTFx	S-1131B39PD-N4YTFz
4.0V±1.0%	S-1131B40UA-N4ZTFx	S-1131B40UC-N4ZTFx	S-1131B40PD-N4ZTFz
4.1V±1.0%	S-1131B41UA-N5ATFx	S-1131B41UC-N5ATFx	S-1131B41PD-N5ATFz
4.2V±1.0%	S-1131B42UA-N5BTFx	S-1131B42UC-N5BTFx	S-1131B42PD-N5BTFz
4.3V±1.0%	S-1131B43UA-N5CTFx	S-1131B43UC-N5CTFx	S-1131B43PD-N5CTFz
4.4V±1.0%	S-1131B44UA-N5DTFx	S-1131B44UC-N5DTFx	S-1131B44PD-N5DTFz
4.5V±1.0%	S-1131B45UA-N5ETFx	S-1131B45UC-N5ETFx	S-1131B45PD-N5ETFz
4.6V±1.0%	S-1131B46UA-N5FTFx	S-1131B46UC-N5FTFx	S-1131B46PD-N5FTFz
4.7V±1.0%	S-1131B47UA-N5GTFx	S-1131B47UC-N5GTFx	S-1131B47PD-N5GTFz
4.8V±1.0%	S-1131B48UA-N5HTFx	S-1131B48UC-N5HTFx	S-1131B48PD-N5HTFz
4.9V±1.0%	S-1131B49UA-N5ITFx	S-1131B49UC-N5ITFx	S-1131B49PD-N5ITFz
5.0V±1.0%	S-1131B50UA-N5JTFx	S-1131B50UC-N5JTFx	S-1131B50PD-N5JTFz
5.1V±1.0%	S-1131B51UA-N5KTFx	S-1131B51UC-N5KTFx	S-1131B51PD-N5KTFz
5.2V±1.0%	S-1131B52UA-N5LTFx	S-1131B52UC-N5LTFx	S-1131B52PD-N5LTFz
5.3V±1.0%	S-1131B53UA-N5MTFx	S-1131B53UC-N5MTFx	S-1131B53PD-N5MTFz
5.4V±1.0%	S-1131B54UA-N5NTFx	S-1131B54UC-N5NTFx	S-1131B54PD-N5NTFz
5.5V±1.0%	S-1131B55UA-N5OTFx	S-1131B55UC-N5OTFx	S-1131B55PD-N5OTFz

**Remark1.** Please contact our sales office for type A products.

- 2. x: G or U
  - z: G or S
- **3.** Please select products of environmental code = U for Sn 100%, halogen-free products.

## ■ Pin Configurations

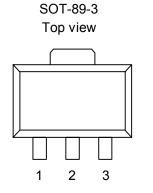


Figure 3



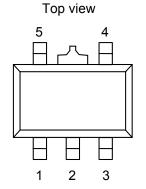


Figure 4



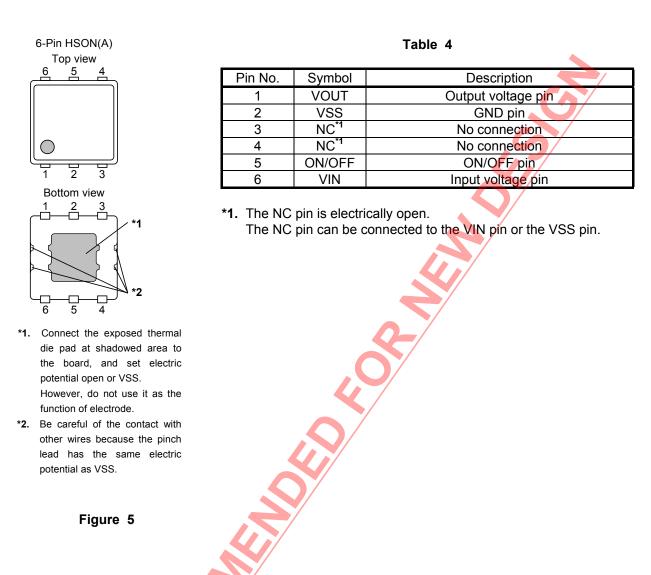
	Description	Symbol	Pin No.
pin	Output voltage pin	VOUT	1
	GND pin	VSS	2
in	Input voltage pin	VIN	3



Pin No.	Symbol	Description
1	VOUT 🧹	Output voltage pin
2	VSS	GND pin
3	NC <sup>*1</sup>	No connection
4	ON/OFF	ON/OFF pin
5	VIN	Input voltage pin

\*1. The NC pin is electrically open. The NC pin can be connected to the VIN pin or the VSS pin.

## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR Rev.4.1\_02 S-1131 Series



## Absolute Maximum Ratings

Table 5

			(Ta = 25°C unless other	wise specified)
Item		Symbol	Absolute Maximum Rating	Unit
		V <sub>IN</sub>	$V_{SS}$ – 0.3 to $V_{SS}$ + 7	V
Input voltage		V <sub>ON/OFF</sub>	$V_{SS}$ – 0.3 to $V_{IN}$ + 0.3	V
Output voltage		V <sub>OUT</sub>	$V_{SS}$ – 0.3 to $V_{IN}$ + 0.3	V
Dowor	SOT-89-3		500	mW
Power dissipation	SOT-89-5	PD	500	mW
uissipation	6-Pin HSON(A)		500	mW
Operation ambient temperature Storage temperature		T <sub>opr</sub>	-40 to +85	О°
		T <sub>stg</sub>	-40 to +125	Ο°

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR S-1131 Series Rev.4.1\_02

Table 6

## Electrical Characteristics

			Table 6 (Ta	= 25°C	unless c	otherwise	e spec	cified)
Item	Symbol	Conditions		Min.	Тур.	Max.	Unit	Test Circuit
Output voltage*1	V <sub>OUT(E)1</sub>	$V_{IN} = V_{OUT(S)} + 1.0$	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, I_{OUT} = 30 \text{ mA}$		V <sub>OUT(S)</sub>	V <sub>ОUT(S)</sub> × 1.01	v	1
	V <sub>OUT(E)2</sub>	$V_{IN} = V_{OUT(S)} + 1.0$	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, I_{OUT} = 80 \text{ mA}$		V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.02	V	1
Output current <sup>*2</sup>	I <sub>OUT</sub>	$V_{IN} \geq V_{OUT(S)} + 1.0$		300 <sup>*5</sup>		—	mA	3
			V <sub>OUT(S)</sub> = 1.5 V		1.00	1.05	V	1
			V <sub>OUT(S)</sub> = 1.6 V		0.90	0.95	V	1
			V <sub>OUT(S)</sub> = 1.7 V		0.80	0.85	V	1
			V <sub>OUT(S)</sub> = 1.8 V		0.70	0.75	V	1
Dropout voltage*3	V <sub>drop</sub>	I <sub>OUT</sub> = 100 mA	V <sub>OUT(S)</sub> = 1.9 V		0.60	0.65	V	1
Diopour voltage	• drop	1001 - 100 11/1	V <sub>OUT(S)</sub> = 2.0 V		0.50	0.60	V	1
			V <sub>OUT(S)</sub> = 2.1 V		0.40	0.55	V	1
			$2.2 \text{ V} \leq \text{V}_{\text{OUT(S)}} \leq 2.5 \text{ V}$	V—	0.30	0.49	V	1
			$2.6 V \le V_{OUT(S)} \le 3.3 V_{OUT(S)}$	/ _	0.25	0.34	V	1
			$3.4 \text{ V} \le \text{V}_{\text{OUT(S)}} \le 5.5 \text{ V}$		0.20	0.28	V	1
Line regulation	$\frac{\Delta V \text{OUT1}}{\Delta V \text{IN} \bullet V \text{OUT}}$	$V_{OUT(S)} + 0.5 V \le V$ $I_{OUT} = 80 \text{ mA}$	$V_{\rm IN} \leq 6.5  \rm V,$	_	0.05	0.2	%/V	1
Load regulation	$\Delta V_{OUT2}$	$V_{IN} = V_{OUT(S)} + 1.0 V,$ 1.0 mA $\leq I_{OUT} \leq 80 \text{ mA}$		_	20	40	mV	1
Output voltage temperature coefficient*4	$\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}$	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, I_{OUT} = 10 \text{ mA}, -40^{\circ}\text{C} \le \text{Ta} \le 85^{\circ}\text{C}$			±100	_	ppm /°C	1
Current consumption during operation	I <sub>SS1</sub>		V, ON/OFF pin = ON,		35	65	μA	2
Input voltage	V <sub>IN</sub>			2.0	_	6.5	V	_
Ripple rejection	RR	$V_{IN} = V_{OUT(S)} + 1.0$ $\Delta V_{rip} = 0.5 Vrms, I$			70	_	dB	5
Short-circuit current	I <sub>short</sub>		V, ON/OFF pin = ON,	_	450	_	mA	3
Current consumption during power-off	I <sub>SS2</sub>		V, ON/OFF pin = OFF,	_	0.1	1.0	μA	2
ON/OFF pin input voltage "H"	V <sub>SH</sub>	$V_{\text{IN}} = V_{\text{OUT}(S)} + 1.0$	V, R <sub>L</sub> = 1.0 kΩ	1.5		_	V	4
ON/OFF pin input voltage "L"	V <sub>SL</sub>	$V_{IN} = V_{OUT(S)} + 1.0$	V, R <sub>L</sub> = 1.0 kΩ			0.3	V	4
ON/OFF pin input current "H"	I <sub>SH</sub>	V <sub>IN</sub> = 6.5 V, V <sub>ON/OF</sub>	<sub>FF</sub> = 6.5 V	-0.1		0.1	μA	4
ON/OFF pin input current "L"	I <sub>SL</sub>	V <sub>IN</sub> = 6.5 V, V <sub>ON/OF</sub>	r <sub>F</sub> = 0 V	-0.1	_	0.1	μA	4

\*1. V<sub>OUT(S)</sub>: Set output voltage

V<sub>OUT(E)1</sub>: Actual output voltage

Output voltage when fixing  $I_{\text{OUT}}(=30\mbox{ mA})$  and inputting  $V_{\text{OUT}(S)}+1.0\mbox{ V}$ 

V<sub>OUT(E)2</sub>: Actual output voltage

Output voltage when fixing  $I_{OUT}$  (= 80 mA) and inputting  $V_{OUT(S)}$  + 1.0 V

\*2. The output current at which the output voltage becomes 95% of VOUT(E)1 after gradually increasing the output current.

\*3. V<sub>drop</sub> = V<sub>IN1</sub> – (V<sub>OUT3</sub> × 0.98)

 $V_{OUT3}$  is the output voltage when  $V_{IN} = V_{OUT(S)} + 1.0$  V and  $I_{OUT} = 100$  mA.

 $V_{IN1}$  is the input voltage at which the output voltage becomes 98% of  $V_{OUT3}$  after gradually decreasing the input voltage. \*4. A change in the temperature of the output voltage [mV/°C] is calculated using the following equation.

$$\frac{\Delta V_{OUT}}{\Delta Ta} [mV/^{\circ}C]^{*1} = V_{OUT(S)} [V]^{*2} \times \frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}} [ppm/^{\circ}C]^{*3} \div 1000$$

\*1. Change in temperature of the output voltage

\*2. Set output voltage

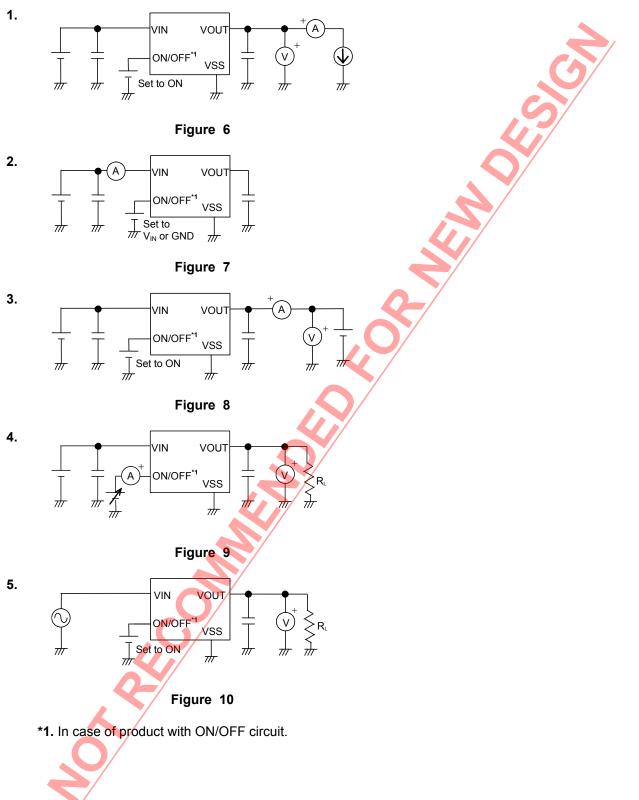
\*3. Output voltage temperature coefficient

**\*5.** The output current can be at least this value.

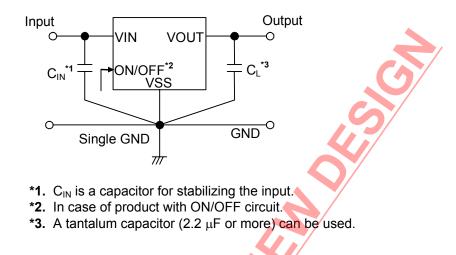
Due to restrictions on the package power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation of the package when the output current is large. This specification is guaranteed by design.

## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR Rev.4.1\_02 S-1131 Series

## Test Circuits



## Standard Circuit



#### Figure 11

Caution The above connection diagram and constant will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constant.

## ■ Condition of Application

Input capacitor ( $C_{IN}$ ): 1.0  $\mu$ F or more Output capacitor ( $C_L$ ): 2.2  $\mu$ F or more (tantalum capacitor)

Caution Generally a series regulator may cause oscillation, depending on the selection of external parts. Check that no oscillation occurs with the application using the above capacitor.

## Explanation of Terms

#### 1. Low dropout voltage regulator

This voltage regulator has the low dropout voltage due to its built-in low on-resistance transistor,

#### 2. Output voltage (VOUT)

The accuracy of the output voltage is ensured at  $\pm 1.0\%$  under the specified conditions of fixed input voltage<sup>\*1</sup>, fixed output current, and fixed temperature.

#### **\*1.** Differs depending the product.

Caution If the above conditions change, the output voltage value may vary and exceed the accuracy range of the output voltage. Refer to "■ Electrical Characteristics" and "■ Characteristics (Typical Data)" for details.

3. Line regulation 
$$\left(\frac{\Delta V_{OUT1}}{\Delta V_{IN} \bullet V_{OUT}}\right)$$

Indicates the dependency of the output voltage on the input voltage. That is, the value shows how much the output voltage changes due to a change in the input voltage with the output current remaining unchanged.

#### 4. Load regulation ( $\Delta V_{OUT2}$ )

Indicates the dependency of the output voltage on the output current. That is, the value shows how much the output voltage changes due to a change in the output current with the input voltage remaining unchanged.

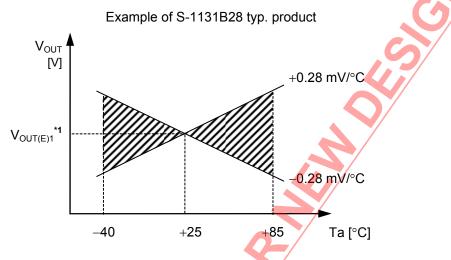
#### 5. Dropout voltage (V<sub>drop</sub>)

Indicates the difference between input voltage ( $V_{IN1}$ ) and the output voltage when; decreasing input voltage ( $V_{IN}$ ) gradually until the output voltage has dropped out to the value of 98% of output voltage ( $V_{OUT3}$ ), which is at  $V_{IN} = V_{OUT(S)} + 1.0 V$ .

$$V_{drop} ~=~ V_{IN1} - (V_{OUT3} \times 0.98)$$

## 6. Output voltage temperature coefficient $\left(\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}\right)$

The shaded area in **Figure 12** is the range where  $V_{OUT}$  varies in the operation temperature range when the output voltage temperature coefficient is  $\pm 100$  ppm/°C.



\*1.  $V_{OUT(E)1}$  is the value of the output voltage measured at Ta = +25°C.

## Figure 12

A change in the temperature of the output voltage [mV/C] is calculated using the following equation.

 $\frac{\Delta V_{OUT}}{\Delta Ta} \left[ mV/^{\circ}C \right]^{*1} = V_{OUT(S)} \left[ V \right]^{*2} \times \frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}} \left[ ppm/^{\circ}C \right]^{*3} \div 1000$ 

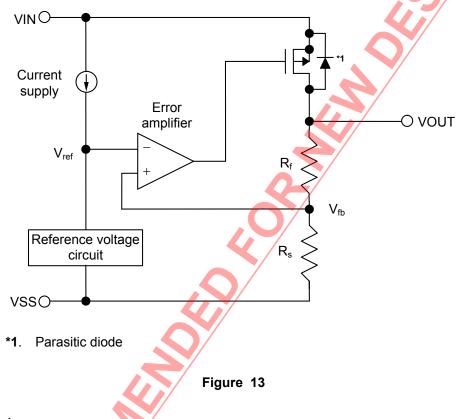
- \*1. Change in temperature of output voltage
- \*2. Set output voltage
- \*3. Output voltage temperature coefficient

## Operation

#### 1. Basic operation

Figure 13 shows the block diagram of the S-1131 Series.

The error amplifier compares the reference voltage ( $V_{ref}$ ) with feedback voltage ( $V_{fb}$ ), which is the output voltage resistance-divided by feedback resistors ( $R_s$  and  $R_f$ ). It supplies the gate voltage necessary to maintain the constant output voltage which is not influenced by the input voltage and temperature change, to the output transistor.



#### 2. Output transistor

In the S-1131 Series, a low on-resistance P-channel MOS FET is used as the output transistor. Be sure that  $V_{OUT}$  does not exceed  $V_{IN} + 0.3$  V to prevent the voltage regulator from being damaged due to reverse current flowing from the VOUT pin through a parasitic diode to the VIN pin, when the potential of  $V_{OUT}$  became higher than  $V_{IN}$ .

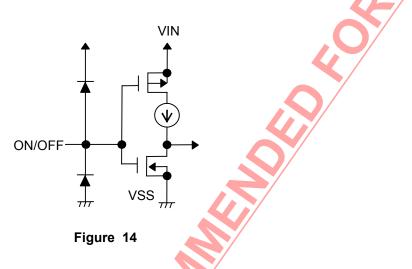
#### 3. ON/OFF pin

In case of product with ON/OFF circuit, this pin starts and stops the regulator.

When the ON/OFF pin is set to OFF level, the entire internal circuit stops operating, and the built-in P-channel MOS FET output transistor between the VIN pin and the VOUT pin is turned off, reducing current consumption significantly. The VOUT pin becomes the Vss level due to the internally divided resistance of several hundreds  $k\Omega$  between the VOUT pin and the VSS pin.

The structure of the ON/OFF pin is as shown in **Figure 14**. Since the ON/OFF pin is neither pulled down nor pulled up internally, do not use it in the floating states. In addition, note that the current consumption increases if a voltage of 0.3 V to  $V_{IN} - 0.3$  V is applied to the ON/OFF pin. When not using the ON/OFF pin, connect it to the VSS pin in the product A type, and connect it to the VIN pin in B type.

Table 7					
	Product Type	ON/OFF Pin	Internal Circuit VOUT Pin Voltage		Current Consumption
	А	"L": ON	Operate	Set value	I <sub>SS1</sub>
	А	"H": OFF	Stop	V <sub>ss</sub> level	I <sub>SS2</sub>
	В	"L": OFF	Stop	V <sub>ss</sub> level	I <sub>SS2</sub>
	В	"H": ON	Operate	Set value	I <sub>SS1</sub>



## ■ Selection of Output Capacitor (C<sub>L</sub>)

The S-1131 Series performs phase compensation using the internal phase compensator in the IC and the ESR (Equivalent Series Resistance) of the output capacitor to enable stable operation independent of changes in the output load. Therefore, always place a capacitor ( $C_L$ ) of 2.2  $\mu$ F or more between the VOUT pin and the VSS pin.

For stable operation of the S-1131 Series, it is essential to employ a capacitor whose ESR is within an optimum range. Using a capacitor whose ESR is outside the optimum range (approximately  $0.5 \Omega$  to  $5 \Omega$ ), whether larger or smaller, may cause an unstable output, resulting in oscillation. For this reason, a tantalum electrolytic capacitor is recommended.

When a ceramic capacitor or an OS capacitor with a low ESR is used, it is necessary to connect an additional resistor that serves as the ESR in series with the output capacitor. The required resistance value is approximately 0.5  $\Omega$  to 5  $\Omega$ , which varies depending on the usage conditions, so perform sufficient evaluation for selection. Ordinarily, around 1.0  $\Omega$  is recommended.

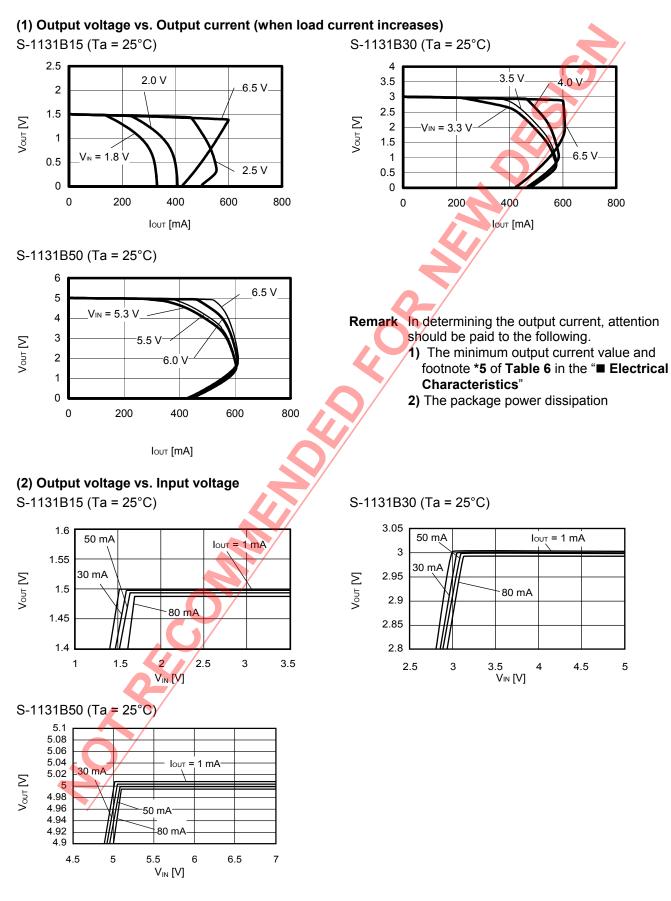
Note that an aluminum electrolytic capacitor may increase the ESR at a low temperature, causing oscillation. When using this kind of capacitor, perform thorough evaluation, including evaluation of temperature characteristics.

### Precautions

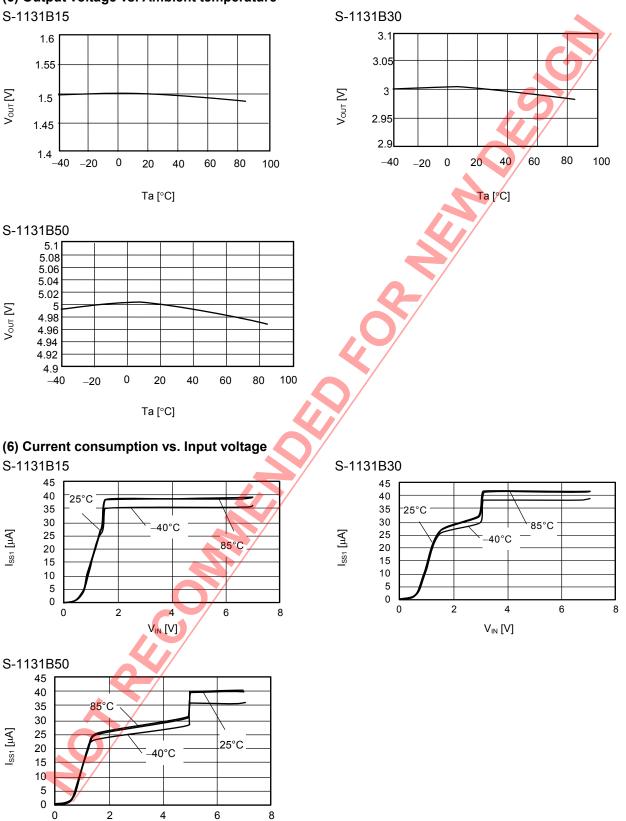
- Wiring patterns for the VIN pin, the VOUT pin and GND should be designed so that the impedance is low. When mounting an output capacitor between the VOUT pin and the VSS pin (C<sub>L</sub>) and a capacitor for stabilizing the input between the VIN pin and the VSS pin (C<sub>IN</sub>), the distance from the capacitors to these pins should be as short as possible.
- Note that generally the output voltage may increase when a series regulator is used at low load current (1.0 mA or less).
- The S-1131 Series performs phase compensation by using an internal phase compensator and the ESR of an output capacitor. Therefore, always place a capacitor of 2.2 μF or more between the VOUT pin and the VSS pin. A tantalum type capacitor is recommended. Moreover, to secure stable operation of the S-1131 Series, it is necessary to employ a capacitor with an ESR within an optimum range (0.5 Ω to 5 Ω). Using a capacitor whose ESR is outside the optimum range (approximately 0.5 Ω to 5 Ω), whether larger or smaller, may cause an unstable output, resulting in oscillation. Perform sufficient evaluation under the actual usage conditions for selection, including evaluation of temperature characteristics.
- The voltage regulator may oscillate when the impedance of the power supply is high and the input capacitance is small or an input capacitor is not connected.
- Overshoot may occur in the output voltage momentarily if the voltage is rapidly raised at power-on or when the power supply fluctuates. Sufficiently evaluate the output voltage at power-on with the actual device.
- The application conditions for the input voltage, the output voltage, and the load current should not exceed the package power dissipation.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- In determining the output current, attention should be paid to the output current value specified in Table 6 in "■ Electrical Characteristics" and footnote \*5 of the table.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.



## ■ Characteristics (Typical Data)



S-1131B15 S-1131B30 1.0 1.0 0.8 0.8 85°C-25°C -85°C 0.6 0.6 Vdrop [V] V<sub>drop</sub> [V] 25°C 0.4 0.4 -40°C 0.2 0.2 40°C 0 ⊾ 0 0 0 50 100 200 250 300 50 100 150 200 250 300 150 350 350 IOUT [mA] lout [mA] S-1131B50 1.0 0.8 0.6 V<sub>drop</sub> [V] 85°Ç 0.4 25°C 0.2 -40°C 0 200 250 300 350 0 50 100 150 IOUT [mA] (4) Dropout voltage vs. Set output voltage 0.4 150 mA\_ 0.35 0.3 100 mA 0.25 V drop [V] 0.2 0.15 50 mA 0.1 30 mA 0.05 ⁻10 mA 0 2 3 5 6 7 0 1 4 Vout(s) [V]



(5) Output voltage vs. Ambient temperature

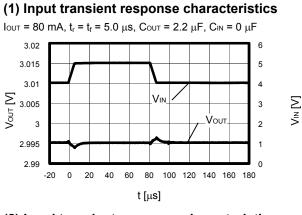
VIN [V]

## HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR Rev.4.1\_02 S-1131 Series

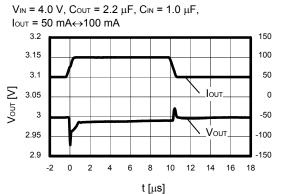
#### (7) Ripple rejection S-1131B15 (Ta = 25°C) S-1131B30 (Ta = 25°C) VIN = 2.5 V, COUT = 2.2 μF V<sub>IN</sub> = 4.0 V, C<sub>OUT</sub> = 2.2 μF 100 100 4 80 Ripple Rejection [dB] 80 Тоит = 1 mA Ripple Rejection [dB] louτ = 1 mA 60 60 40 40 30 mA 30 mA 20 20 80 mA 80 mA 0 0 10 100 1 k 10 k 100 k 10 100 10 k 100 k 1 M 1 M 1 k Frequency [Hz] Frequency [Hz] S-1131B50 (Ta = 25°C) $V_{IN}$ = 6.0 V, Cout = 2.2 $\mu$ F 100 80 Ripple Rejection [dB] Тоит = 1 mA 60 40 30 mA 20 80 mA 0 10 100 1 k 10 k 100 k 1 M Frequency [Hz]

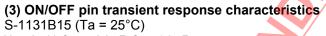
lour [mA]

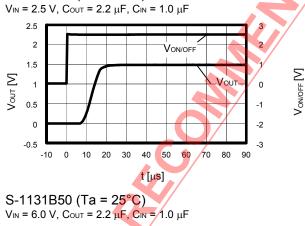
## Reference Data

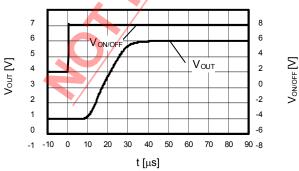


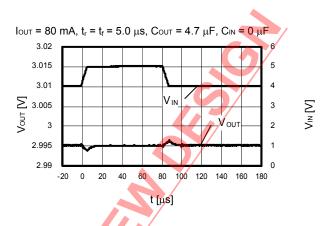
#### (2) Load transient response characteristics

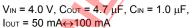


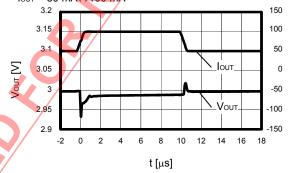






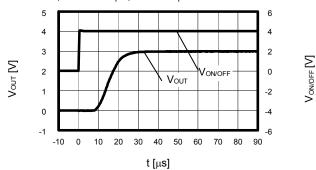


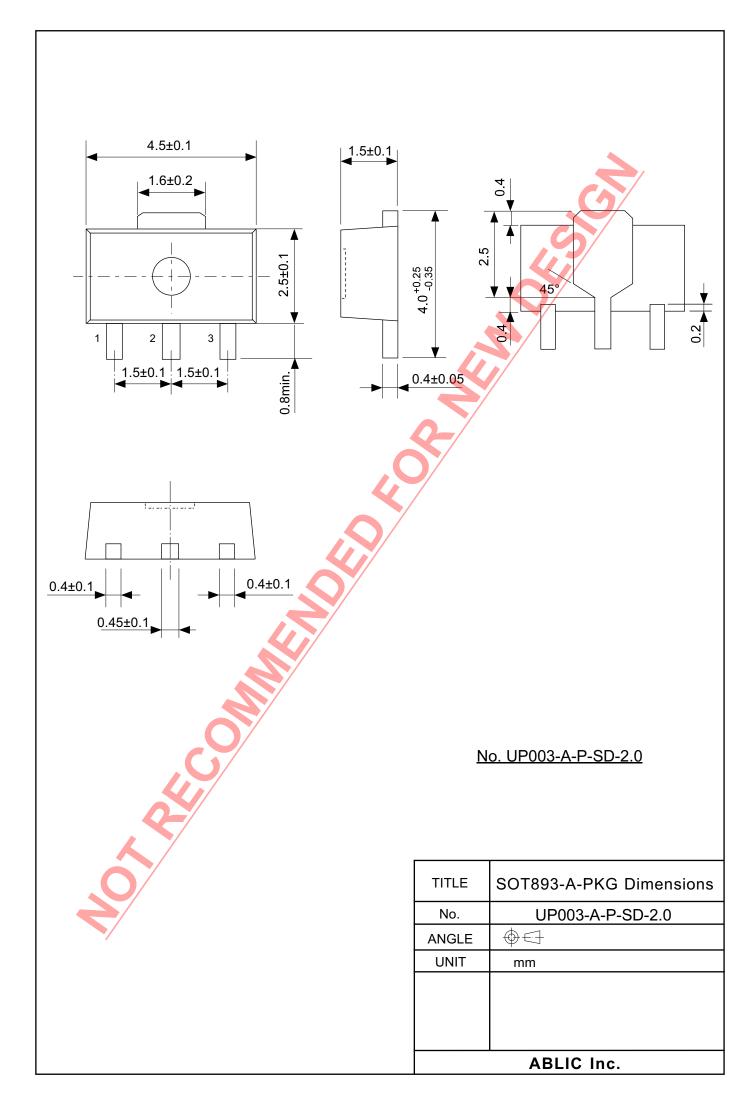


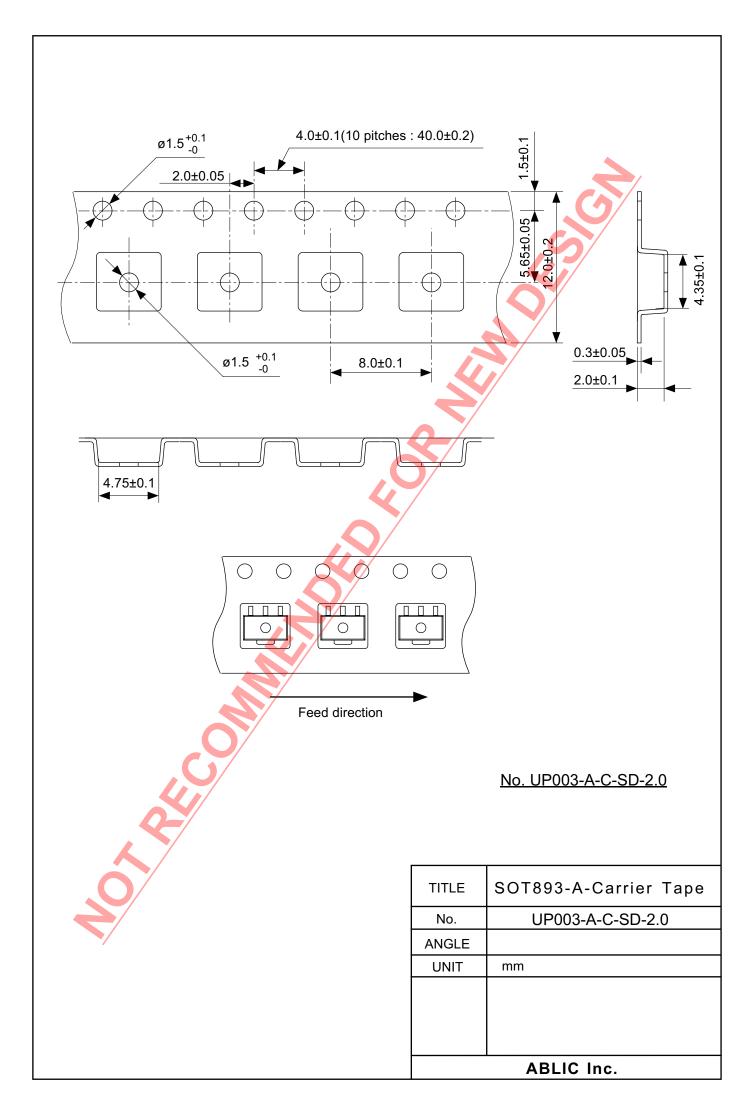


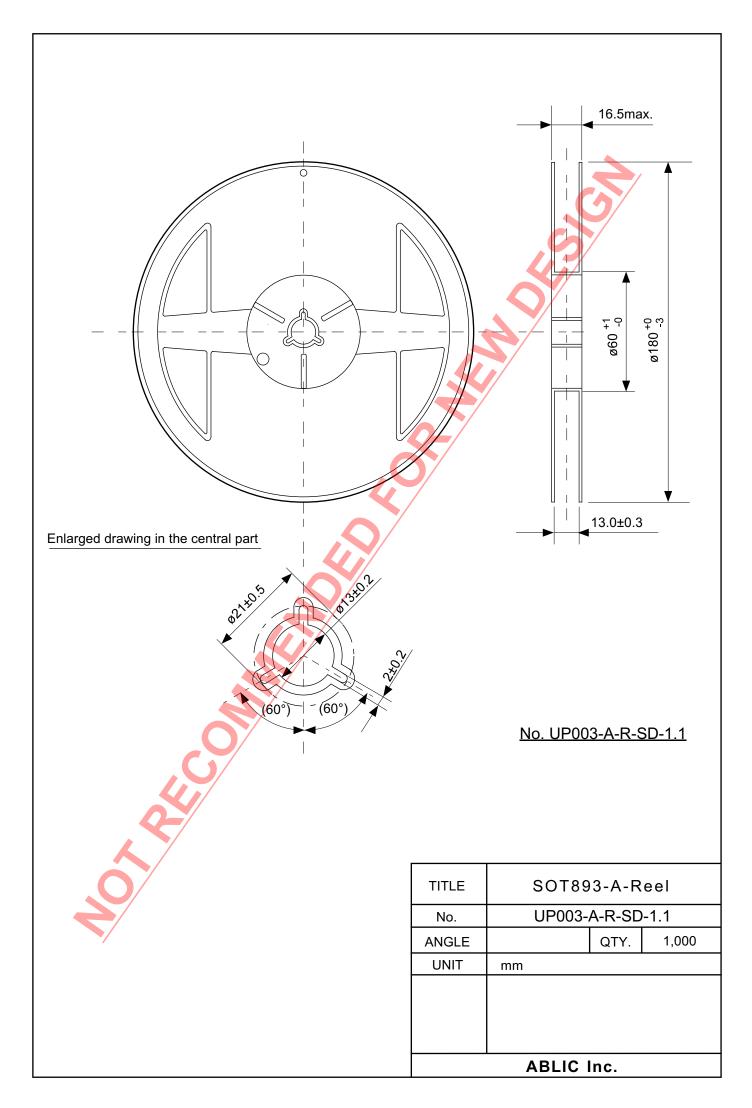
lour [mA]

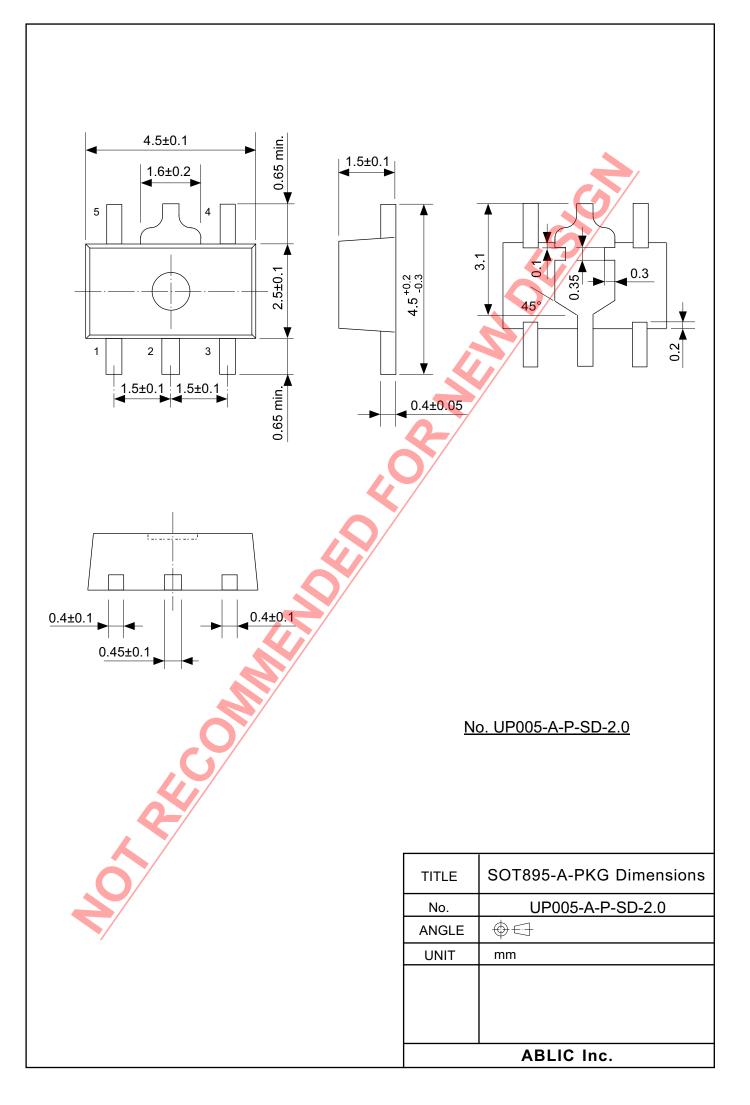
S-1131B30 (Ta = 25°C) ViN = 4.0 V, Cout = 2.2 μF, CiN = 1.0 μF

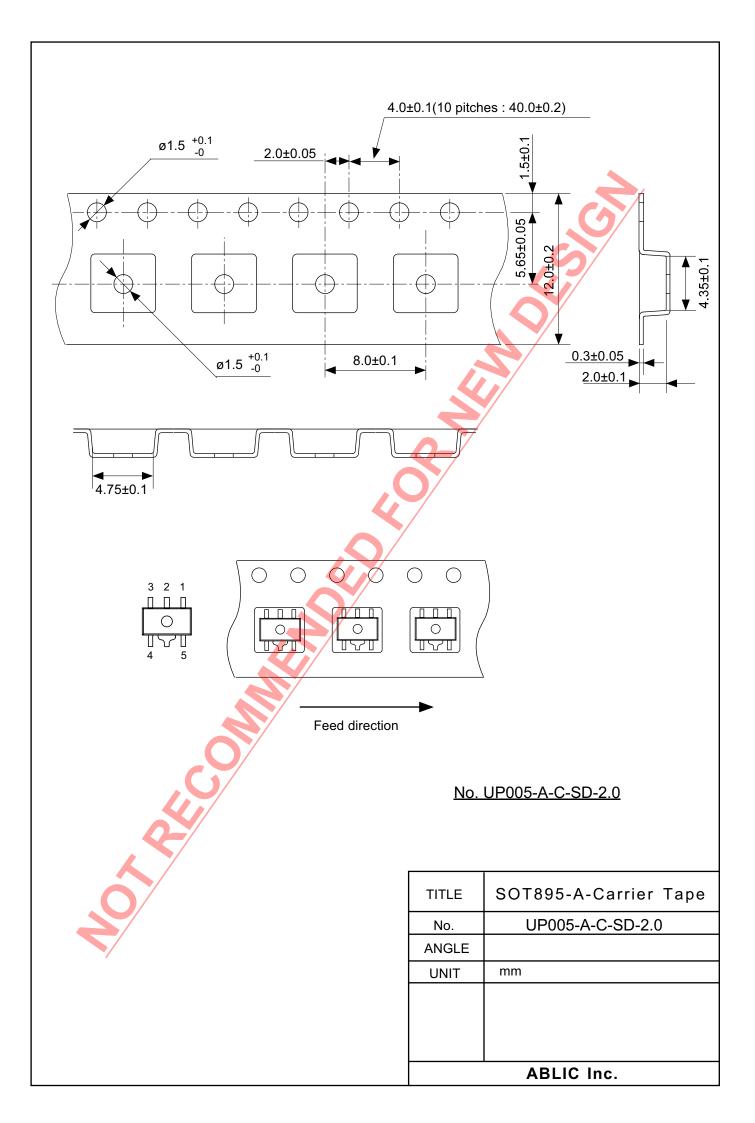


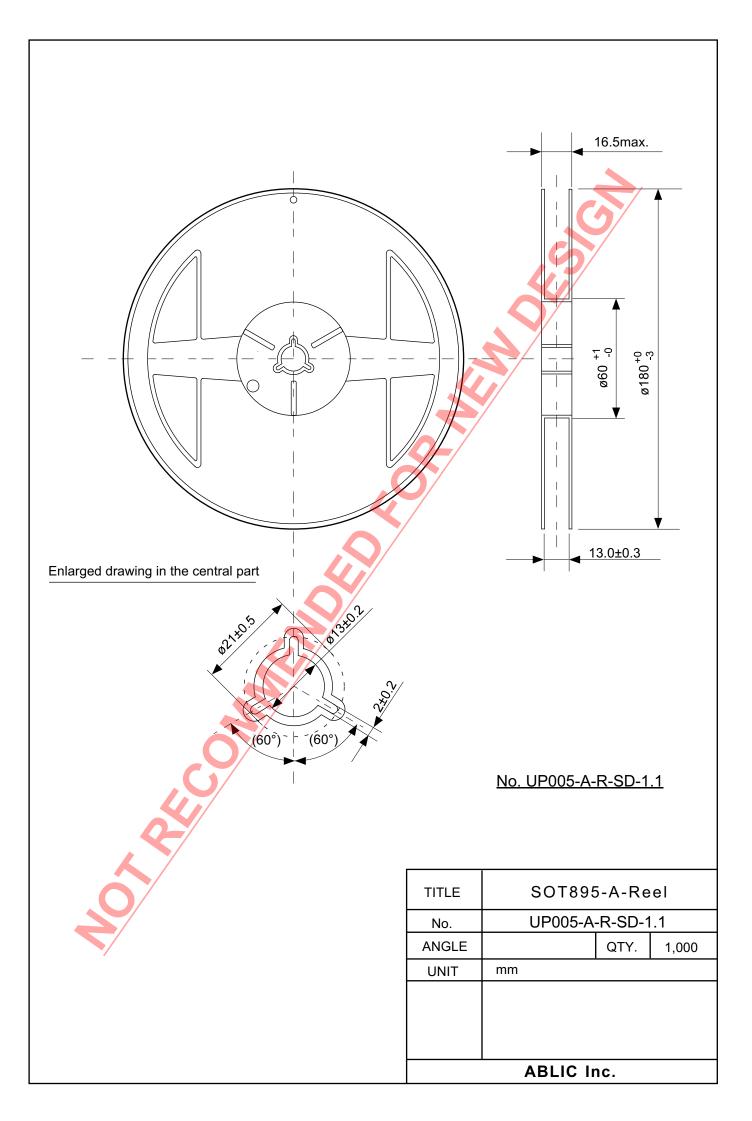


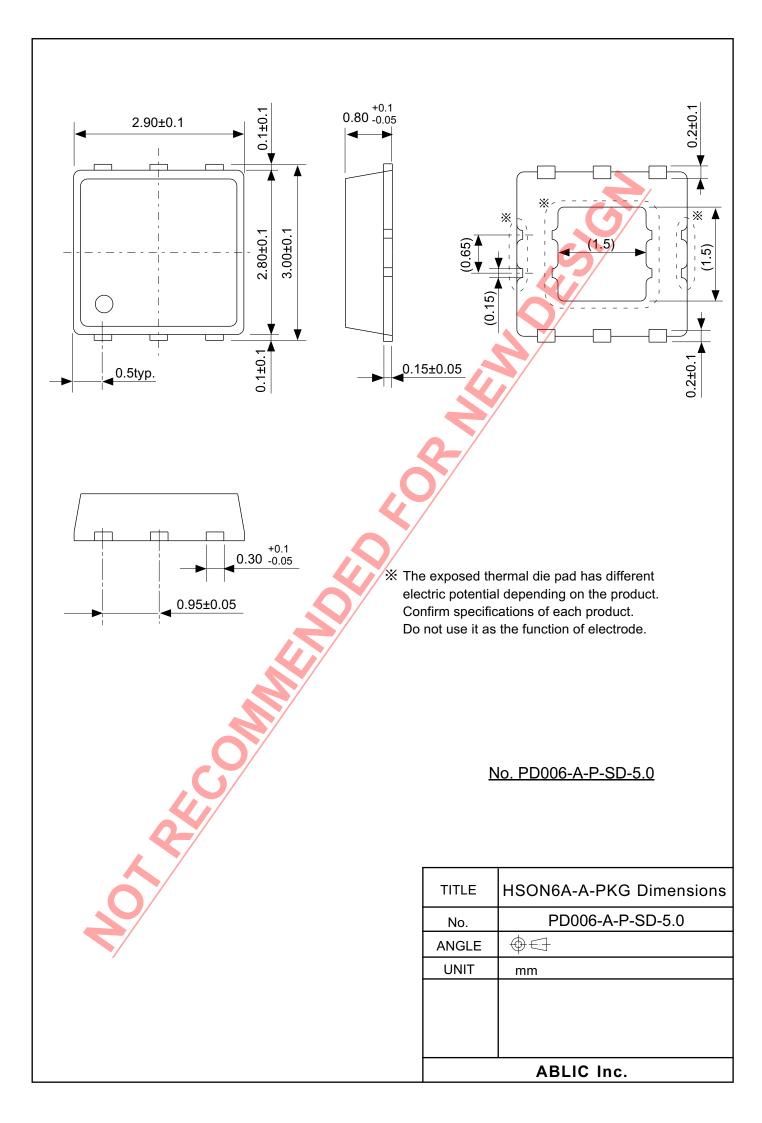


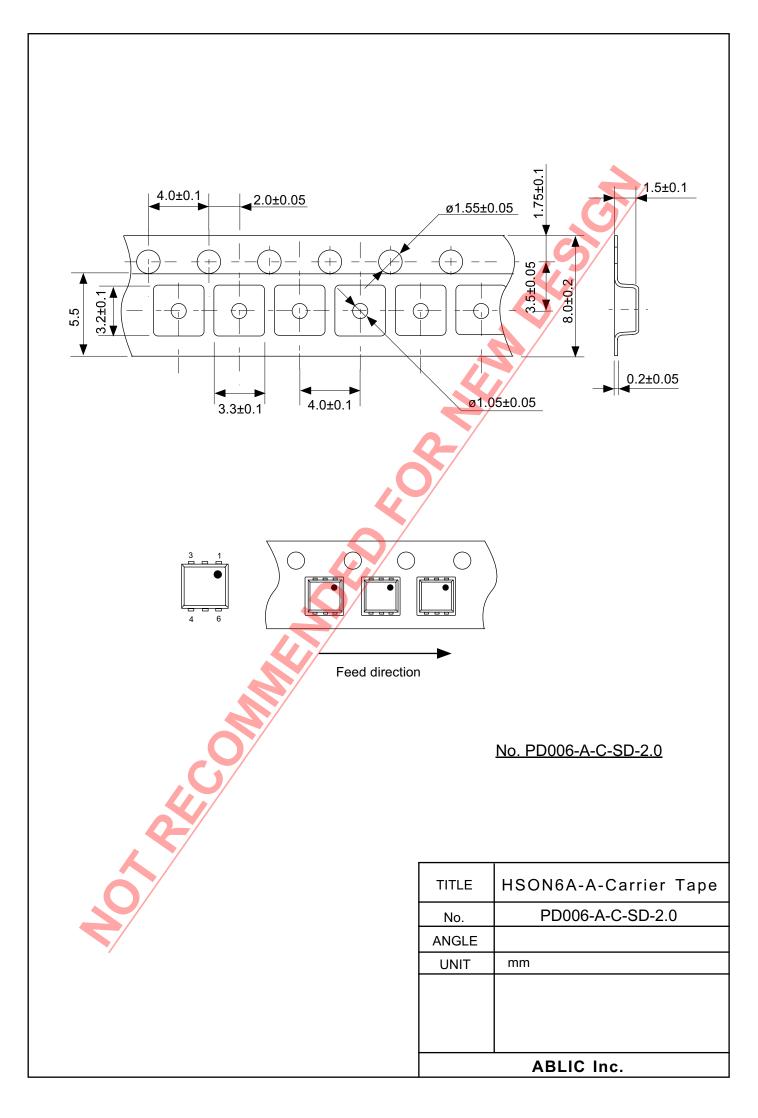


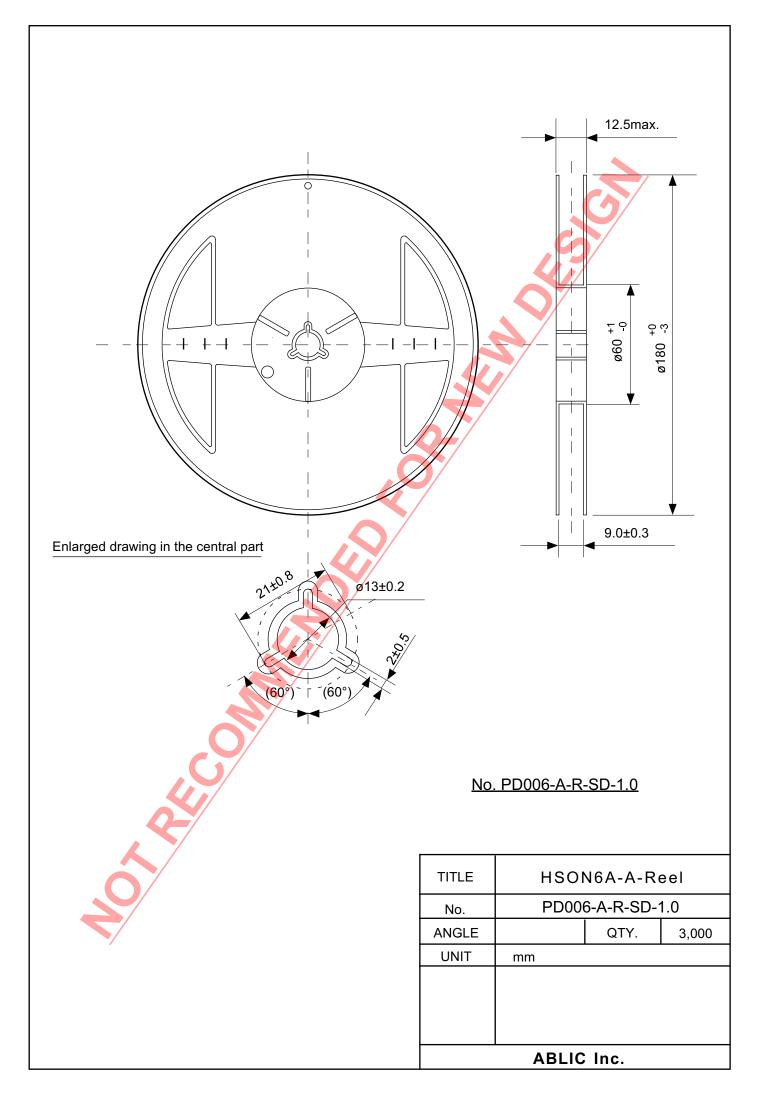












## **Disclaimers (Handling Precautions)**

- 1. All the information described herein (product data, specifications, figures, tables, programs, algorithms and application circuit examples, etc.) is current as of publishing date of this document and is subject to change without notice.
- The circuit examples and the usages described herein are for reference only, and do not guarantee the success of any specific mass-production design.
   ABLIC Inc. is not liable for any losses, damages, claims or demands caused by the reasons other than the products described herein (hereinafter "the products") or infringement of third-party intellectual property right and any other right due to the use of the information described herein.
- 3. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by the incorrect information described herein.
- 4. Be careful to use the products within their ranges described herein. Pay special attention for use to the absolute maximum ratings, operation voltage range and electrical characteristics, etc. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by failures and/or accidents, etc. due to the use of the products outside their specified ranges.
- 5. Before using the products, confirm their applications, and the laws and regulations of the region or country where they are used and verify suitability, safety and other factors for the intended use.
- 6. When exporting the products, comply with the Foreign Exchange and Foreign Trade Act and all other export-related laws, and follow the required procedures.
- 7. The products are strictly prohibited from using, providing or exporting for the purposes of the development of weapons of mass destruction or military use. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by any provision or export to the person or entity who intends to develop, manufacture, use or store nuclear, biological or chemical weapons or missiles, or use any other military purposes.
- 8. The products are not designed to be used as part of any device or equipment that may affect the human body, human life, or assets (such as medical equipment, disaster prevention systems, security systems, combustion control systems, infrastructure control systems, vehicle equipment, traffic systems, in-vehicle equipment, aviation equipment, aerospace equipment, and nuclear-related equipment), excluding when specified for in-vehicle use or other uses by ABLIC, Inc. Do not apply the products to the above listed devices and equipments. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by unauthorized or unspecified use of the products.
- 9. In general, semiconductor products may fail or malfunction with some probability. The user of the products should therefore take responsibility to give thorough consideration to safety design including redundancy, fire spread prevention measures, and malfunction prevention to prevent accidents causing injury or death, fires and social damage, etc. that may ensue from the products' failure or malfunction.

The entire system in which the products are used must be sufficiently evaluated and judged whether the products are allowed to apply for the system on customer's own responsibility.

- 10. The products are not designed to be radiation-proof. The necessary radiation measures should be taken in the product design by the customer depending on the intended use.
- 11. The products do not affect human health under normal use. However, they contain chemical substances and heavy metals and should therefore not be put in the mouth. The fracture surfaces of wafers and chips may be sharp. Be careful when handling these with the bare hands to prevent injuries, etc.
- 12. When disposing of the products, comply with the laws and ordinances of the country or region where they are used.
- 13. The information described herein contains copyright information and know-how of ABLIC Inc. The information described herein does not convey any license under any intellectual property rights or any other rights belonging to ABLIC Inc. or a third party. Reproduction or copying of the information from this document or any part of this document described herein for the purpose of disclosing it to a third-party is strictly prohibited without the express permission of ABLIC Inc.
- 14. For more details on the information described herein or any other questions, please contact ABLIC Inc.'s sales representative.
- 15. This Disclaimers have been delivered in a text using the Japanese language, which text, despite any translations into the English language and the Chinese language, shall be controlling.



2.4-2019.07